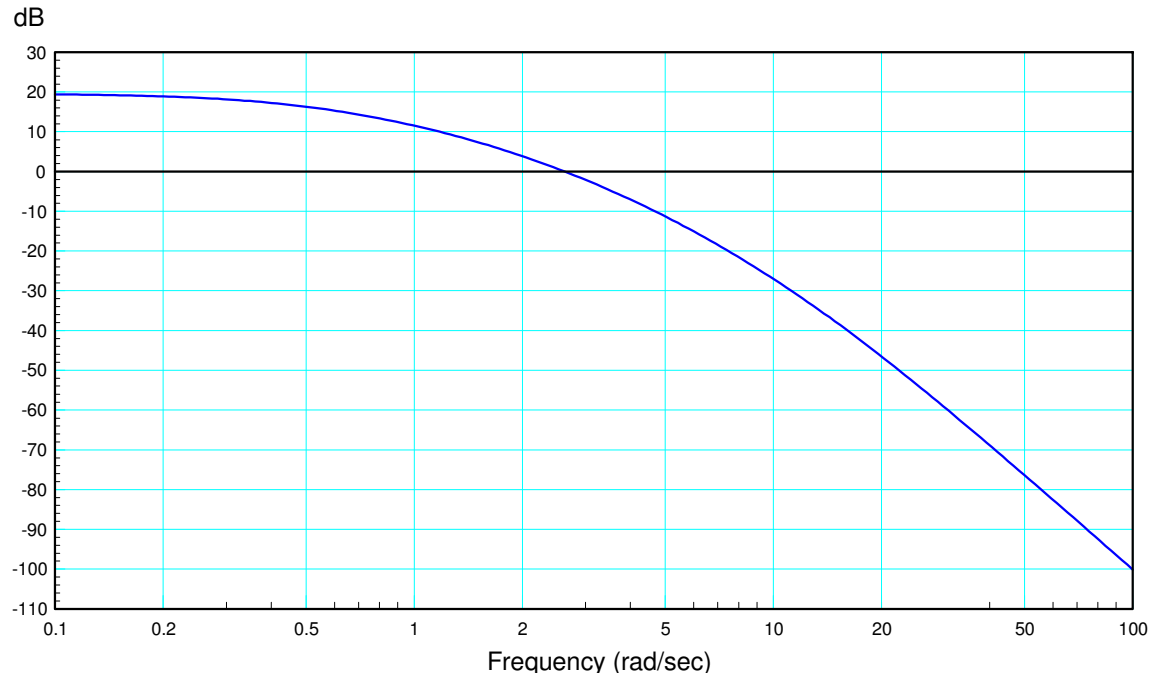


Homework #12: ECE 461/661

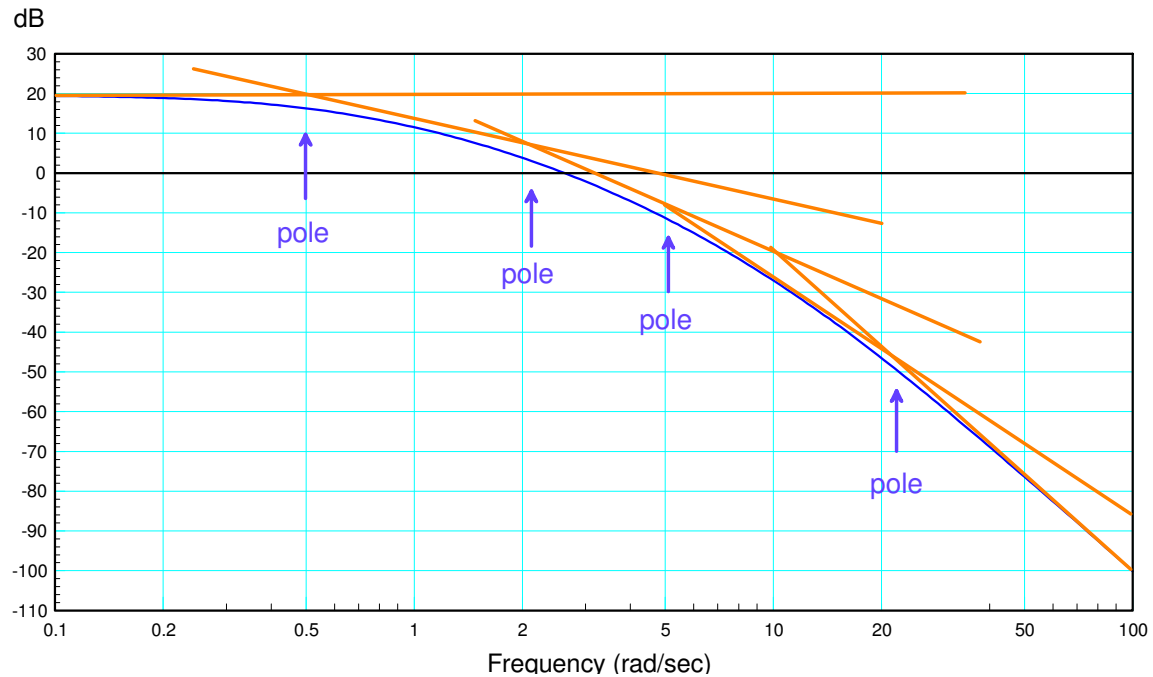
Bode Plots, Nichols charts and gain & lead compensation. Due Monday, December 2nd

Bode Plots

1) Determine the system, $G(s)$, with the following gain vs. frequency



Step 1: Draw in the straight-line asymptotes. Make sure each is at a multiple of 20dB/decade



Each corner (where the asymptotes meet) is a pole

$$G(s) = \left(\frac{k}{(s+0.5)(s+2.1)(s+5)(s+21)} \right)$$

Pick 'k' to match the DC gain (approx the same as the gain at $s = j0.1$)

$$\left(\frac{k}{(s+0.5)(s+2.1)(s+5)(s+21)} \right)_{s=0} = 20dB = 10$$

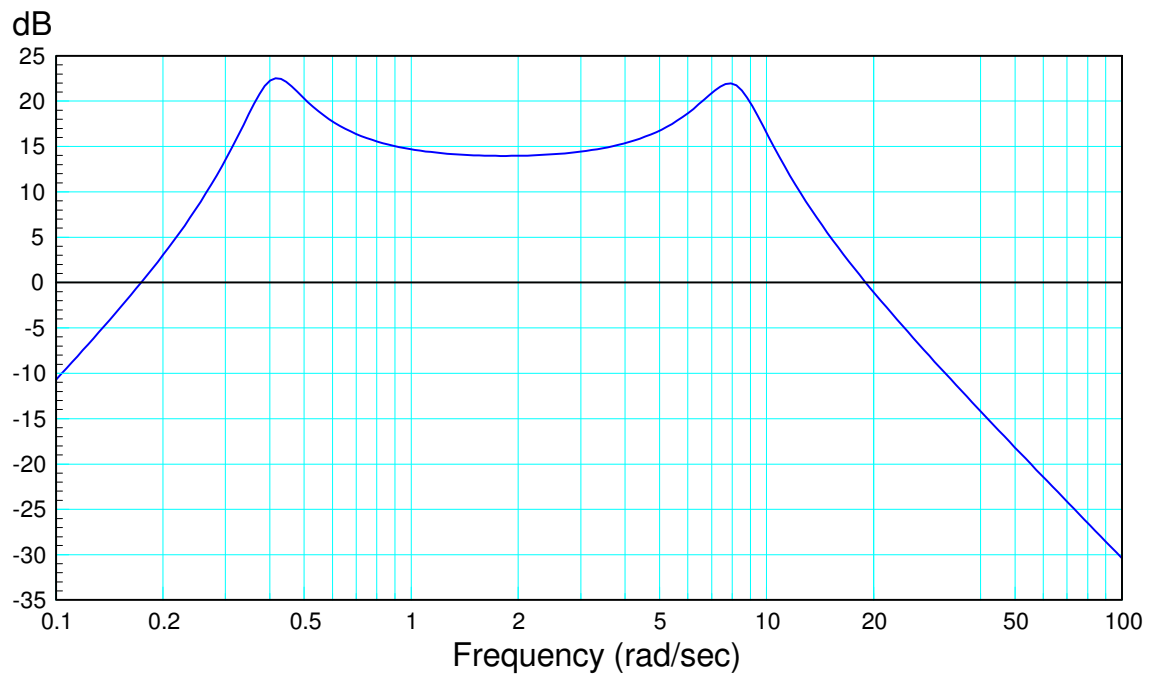
$$k = 1102.5$$

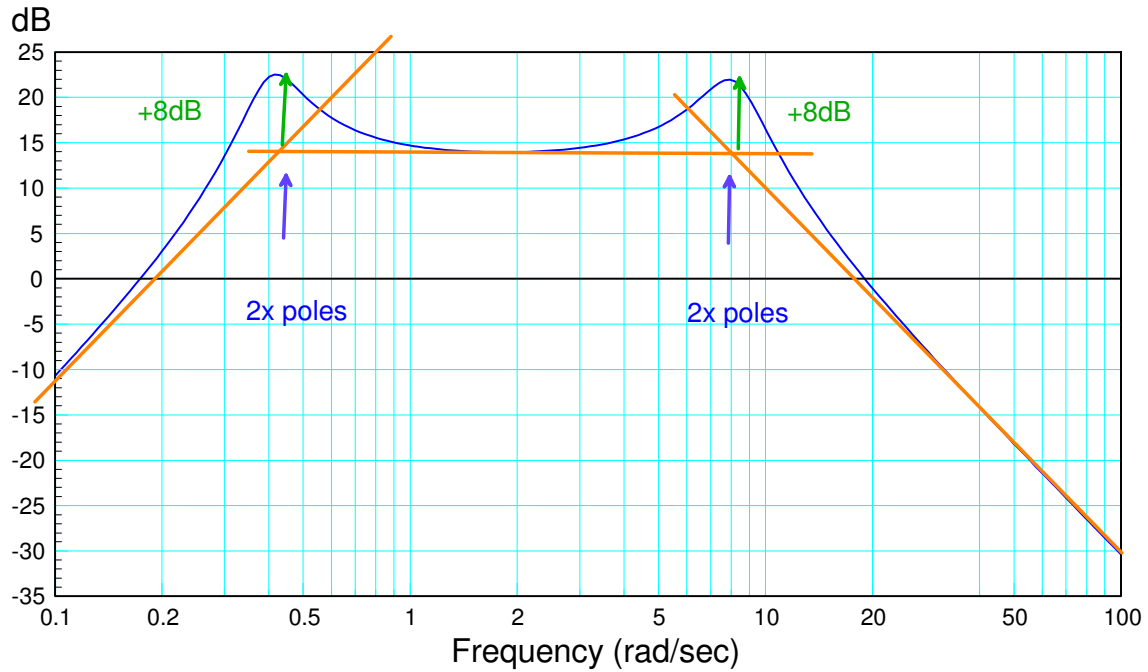
so

$$G(s) \approx \left(\frac{1102.5}{(s+0.5)(s+2.1)(s+5)(s+21)} \right)$$

answers will vary

2) Determine the system, $G(s)$, with the following gain vs. frequency





Draw in the straight-line asymptotes at multiples of 20dB/decade

- shown on graph

The corners tell you the magnitude of the poles & zeros

- Two zeros left of 0.1 (assume at $s = 0$)
- Two poles at 0.42 rad/sec
- Two poles at 8 rad/sec

$$G(s) \approx \left(\frac{ks^2}{(s+0.42\angle\pm\theta)(s+8\angle\pm\phi)} \right)$$

The gain at the corner tells you the damping ratio

$$\text{gain at corner} = +8\text{dB} = 2.5119$$

$$\frac{1}{2\zeta} = 2.5119$$

$$\zeta = 0.199$$

$$\theta = 78.52^\circ$$

Same for ϕ

To find k, pick k to match the gain somewhere, like $s = j2$

$$\left(\frac{ks^2}{(s+0.42\angle\pm 78.52^\circ)(s+8\angle\pm 78.52^\circ)} \right)_{s=j2} = 14dB = 5.012$$

$$k =$$

In Matlab:

```
>> p1 = -0.42 * exp(j*78.52*pi/180);
>> p2 = conj(p1);
>> p3 = -8*exp(j*78.52*pi/180);
>> p4 = conj(p3);
>> G = zpk([0,0],[p1,p2,p3,p4],1);
>> evalfr(G,j*2)
```

```
ans = 0.0173 - 0.0003i
```

```
>> k = 5.012 / abs(ans)
```

```
k = 290.1763
```

so

$$G(s) \approx \left(\frac{290.1763s^2}{(s+0.42\angle\pm 78.52^\circ)(s+8\angle\pm 78.52^\circ)} \right)$$

Checking in Matlab

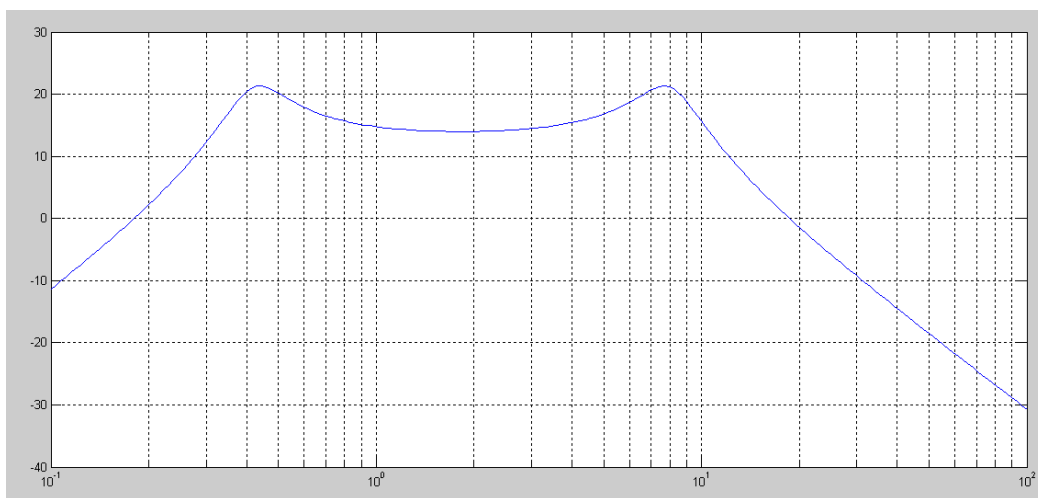
```
>> G = zpk([0,0],[-p1,-p2,-p3,-p4],290.1763)
```

Zero/pole/gain:

```
290.1763 s^2
```

```
-----
(s^2 + 0.1672s + 0.1764) (s^2 + 3.184s + 64)
```

```
>> Gw = Bode2(G,w);
>> semilogx(w,20*log10(abs(Gw)));
>> grid on
```



Nichols Charts

3) The gain vs. frequency of a system is measured

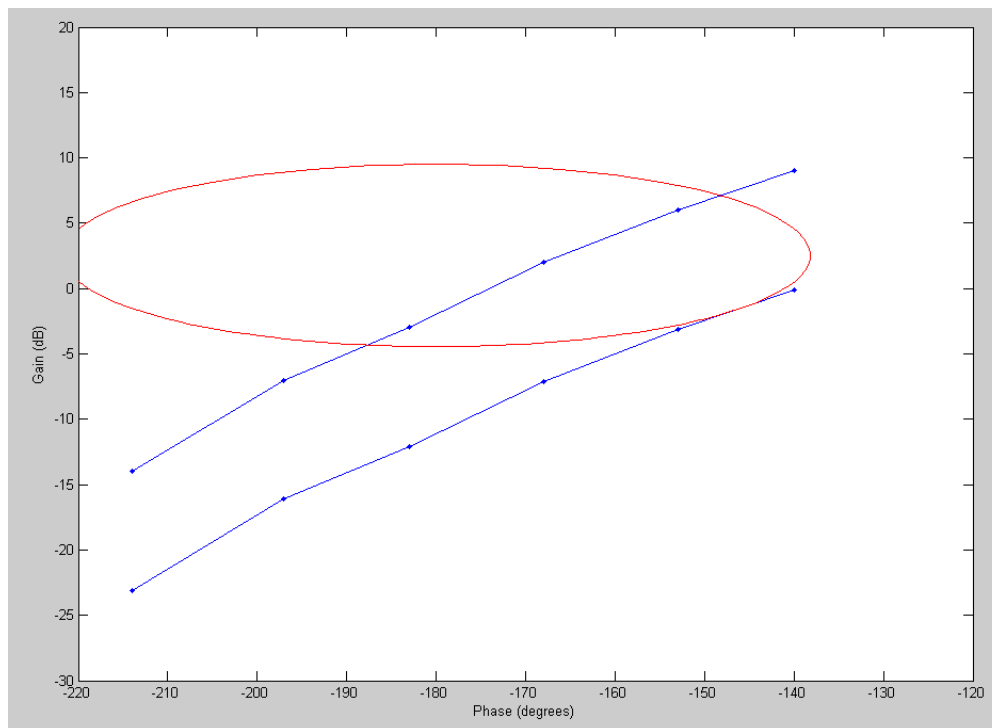
w (rad/sec)	2	3	4	5	6	10
Gain (dB)	9	6	2	-3	-7	-14
Phase (deg)	-140	-153	-168	-183	-197	-214

Using this data

- Transfer it to a Nichols chart
- Determine the maximum gain that results in a stable system
- Determine the gain, k , that results in a maximum closed-loop gain of $M_m = 1.5$

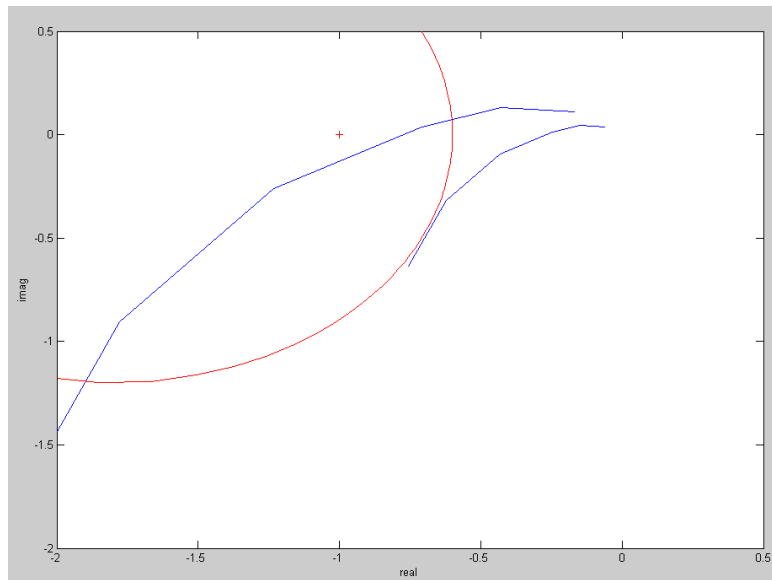
```
>> w = [2, 3, 4, 5, 6, 10]';  
>> GdB = [9, 6, 2, -3, -7, -14]';  
>> Gp = [-140, -153, -168, -183, -197, -214]';  
>> Gw = 10 .^(GdB/20) .* exp(j*Gp*pi/180)  
>> nichols2(Gw*[1, 0.5], 1.5);  
>> nichols2(Gw*[1, 0.3], 1.5);  
>> nichols2(Gw*[1, 0.34], 1.5);  
>> nichols2(Gw*[1, 0.35], 1.5);  
>>
```

ans: $k = 0.35$



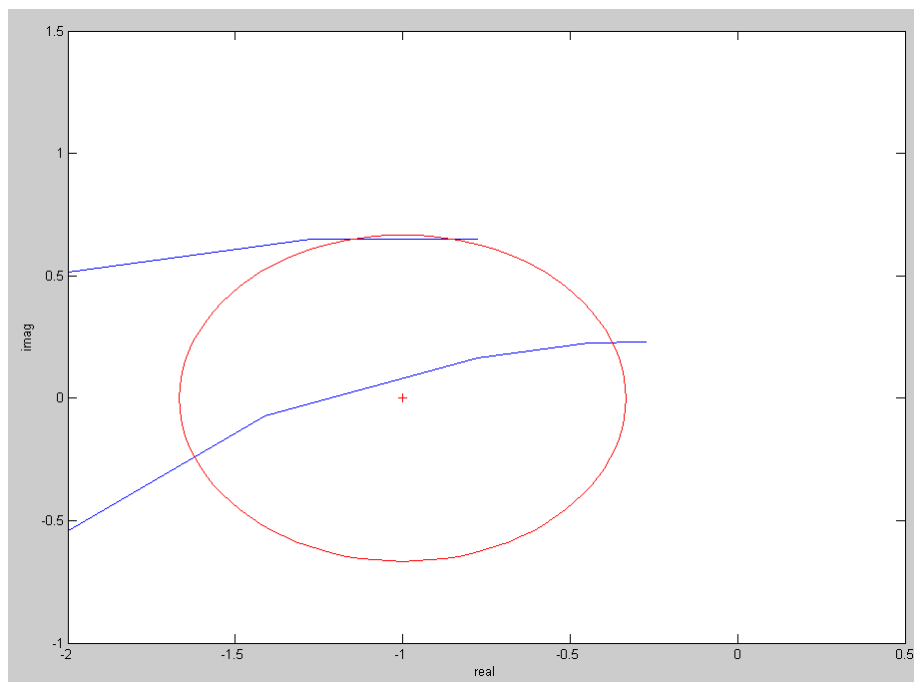
Sidelight: It's the same procedure with a Nyquist diagram

```
>> Nyquist2(Gw*[1,0.35],1.5);
```



or inverse-Nyquist diagram

```
InverseNyquist2(Gw*[1,0.35],1.5);
```



it would look better with more points (take more measurements or do a spline curve fit to add more points)

Gain and Lead Compensation

Problem 4 & 5) Assume

$$G(s) = \left(\frac{300}{s(s+3)(s+8)(s+12)} \right)$$

4) Design a gain compensator that results in a 40 degree phase margin.

- Check the resulting step response in Matlab

For a 40 degree phase margin, at some frequency

$$GK(j\omega) = 1 \angle -140^\circ$$

Step 1: Find where $G(j\omega) = -140$ degrees

$$G(j1.7155) = 0.5120 \angle -140^\circ$$

so

$$k = \frac{1}{0.5120} = 1.9599$$

This should results in...

Phase margin = 40 degrees

$$M_m \approx \left(\frac{1 \angle -140^\circ}{1 + 1 \angle -140^\circ} \right) = 1.4619$$

$$M_m = \frac{1}{2\zeta\sqrt{1-\zeta^2}}$$

$$\zeta = 0.3678$$

$$OS = \exp\left(\frac{-\pi\zeta}{\sqrt{1-\zeta^2}}\right) = 28.86\%$$

The closed-loop dominant pole is approximately:

$$s = -a + jb$$

$$b = 1.7155 \quad \text{the complex part of the pole is approximately the resonance freq}$$

$$a = \frac{b}{\tan(\arccos(\zeta))} = 0.6785$$

$$T_s = \frac{4}{a} = 5.89 \text{ seconds}$$

Ckecking in Matlab:

```
>> G = zpk([], [0, -3, -8, -12], 300)

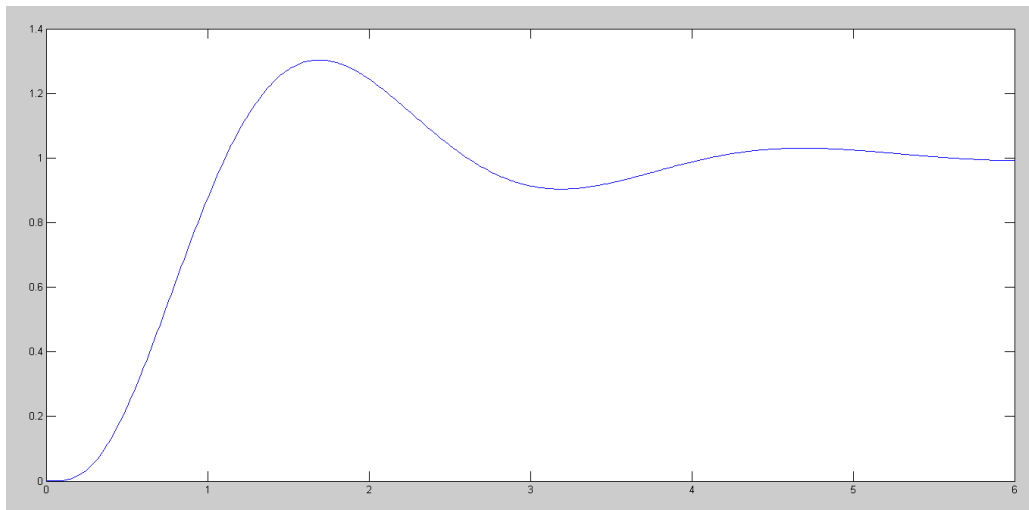
      300
-----
s (s+3) (s+8) (s+12)

>> k = 1.9599;
>> t = [0:0.01:6]';

>> Gcl = minreal(G*k / (1+G*k));
>> y = step(Gcl, t);
>> max(y)

ans =    1.3033

>> plot(t, y)
```



30.33% overshoot

- designed for 28.86% overshoot
- designing for a phase margin is usually slight aggressive

2% settling time is about 5.89 seconds

5) Design a lead compensator that results in a 40 degree phase margin.

- Check the resulting step response in Matlab

$$G(s) = \left(\frac{300}{s(s+3)(s+8)(s+12)} \right)$$

Let

$$K(s) = k \left(\frac{s+3}{s+30} \right)$$

Find the frequency where the phase of GK = -140 degrees

$$GK = \left(\frac{300k}{s(s+8)(s+12)(s+30)} \right)$$

$$GK(j3.7836) = 0.0235k \angle -140^\circ$$

Pick k to make the gain one at this frequency

$$k = \frac{1}{0.0235} = 42.46$$

so

$$K(s) = 42.46 \left(\frac{s+3}{s+30} \right)$$

The closed-loop response should have

- 28.86% overshoot (same phase margin)

$$s = -a + jb$$

$$b = 3.7876$$

$$a = \frac{b}{\tan(\arccos(\zeta))} = 1.4981$$

$$T_s = \frac{4}{1.4981} = 2.67 \text{ seconds}$$

Checking in Matlab

```
>> G = zpk([], [0, -3, -8, -12], 300)
```

$$\frac{300}{s(s+3)(s+8)(s+12)}$$

```
>> K = zpk(-3, -30, 42.46)
```

$$\frac{42.46(s+3)}{(s+30)}$$

```
>> Gcl = minreal(G*K / (1+G*K));
```

```
>> t = [0:0.01:6]';
```

```
>> y = step(Gcl, t);
```

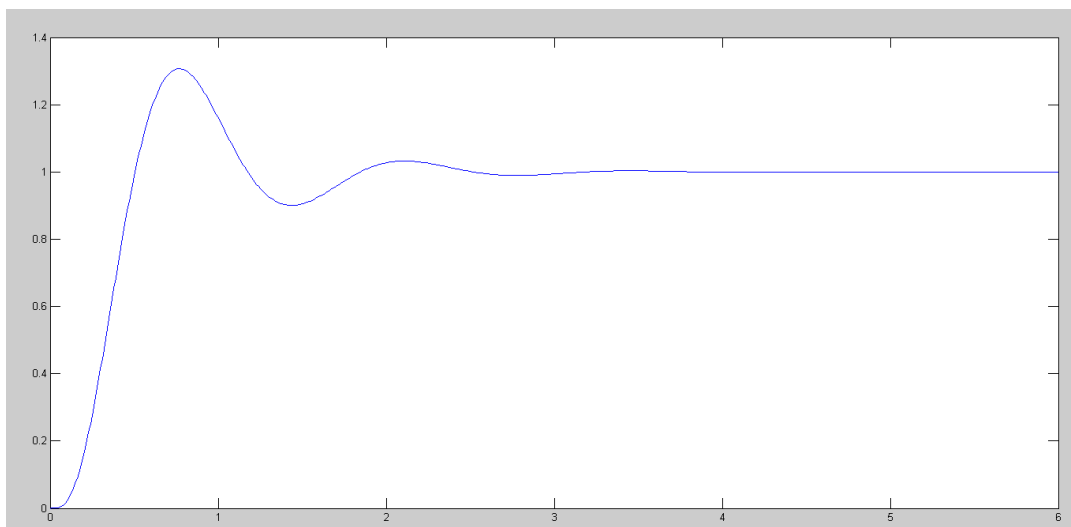
```
>> max(y)
```

```
1.3065
```

```
>> plot(t, y)
```

The step response is about as expected:

- Settling time is about 2.67 seconds as expected
- Overshoot is also about 28.86%



Problem 6 & 7) Assume a 200ms delay is added

$$G(s) = \left(\frac{300}{s(s+3)(s+8)(s+12)} \right) e^{-0.2s}$$

6) Design a gain compensator that results in a 40 degree phase margin.

- Check the resulting step response in Matlab

Same procedure as before. Search along the $j\omega$ axis until the phase is -140 degrees

$$G(j1.2052) = 0.7891 \angle -140^\circ$$

Pick k to make the gain one

$$k = \frac{1}{0.7891} = 1.2673$$

This should result in

- $OS = 28.88\%$ (same phase margin as before)
- $real(s) = \frac{1.2052}{\tan(\arccos(\zeta))} = 0.4767$
- $T_s = \frac{4}{0.4767} = 8.39 \text{ sec}$

Checking in Matlab (use a Pade approximation for a delay)

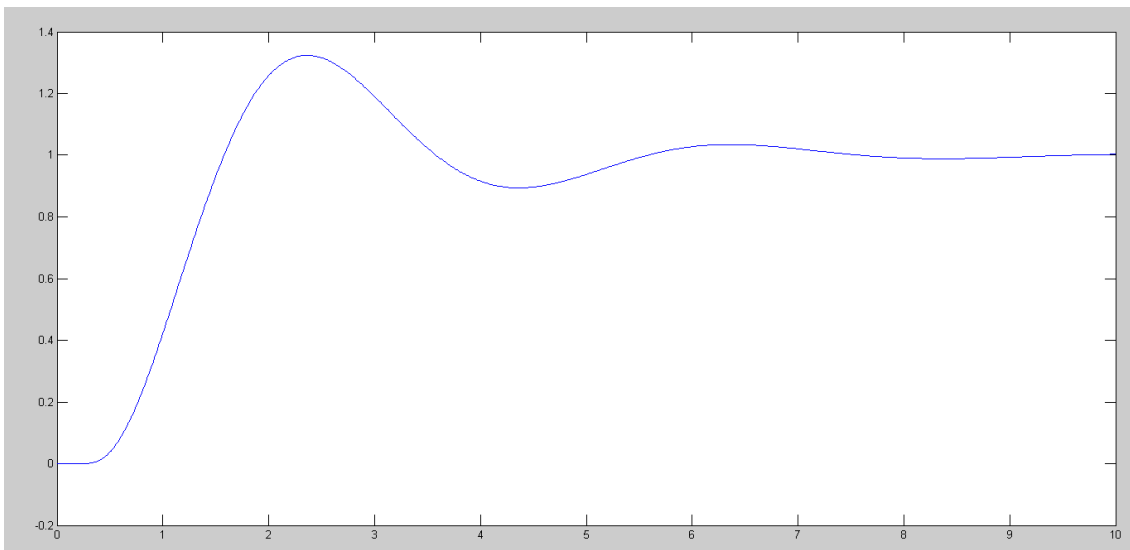
```
>> G = zpke([], [0, -3, -8, -12], 300)

Zero/pole/gain:
      300
-----
s (s+3) (s+8) (s+12)

>> [num,den] = pade(0.2, 6);
>> D = tf(num,den);
>> k = 1.2673;
>> Gcl = minreal(G*D*k / (1+G*D*k));
>> t = [0:0.01:10]';
>> y = step(Gcl, t);
>> max(y)

ans =     1.3235

>> plot(t,y)
>>
```



The results are about as expected

- 32.3% overshoot (vs. 28.8%)
- T_s is about 8.39 seconds

Note: Delays are not an issue when working in the frequency domain (unlike root-locus)

7) Design a lead compensator that results in a 40 degree phase margin.

- Check the resulting step response in Matlab

$$G(s) = \left(\frac{300}{s(s+3)(s+8)(s+12)} \right) e^{-0.2s}$$

Let

$$K(s) = k \left(\frac{s+3}{s+30} \right)$$

so

$$GK = \left(\frac{300k}{s(s+8)(s+12)(s+30)} \right) e^{-0.2s}$$

Search along the jw axis until the phase is -140 degrees

$$G(j1.9907) = 0.05 \angle -140^\circ$$

meaning

$$k = \frac{1}{0.05} = 20.01$$

and

$$K(s) = 20.01 \left(\frac{s+3}{s+30} \right)$$

This should result in

28.88% overshoot (40 degree phase margin)

$$\text{real}(s) = \frac{1.9907}{\tan(\arccos(\zeta))} = 0.7874$$

$$T_s = \frac{4}{0.7874} = 5.08 \text{ seconds}$$

Checking in Matlab

```

>> G = zpk([], [0, -3, -8, -12], 300)

          300
-----
s (s+3) (s+8) (s+12)

>> K = zpk(-3, -30, 20.01)

20.01 (s+3)
-----
(s+30)

>> Gcl = minreal(G*D*K / (1+G*D*K));
>> t = [0:0.01:10]';
>> y = step(Gcl, t);
>> max(y)

ans =    1.3434

>> plot(t, y)

```

The results are about as expected

- 34.34% overshoot (vs. 28.88% expected)
- Ts is about 5.08 seconds
- The delay forces you to back off on the gain giving you a slower system

