

# Homework #11: ECE 461/661

Digital PID Control. Due Monday, November 18th

## PID Control

Assume  $T = 0.1$  seconds:

$$G(s) = \left( \frac{2331}{(s+2.6338)(s+30.2062)(s+53.7896)} \right)$$

1) Design a digital I controller

$$K(z) = k \left( \frac{z}{z-1} \right)$$

that results in 20% overshoot in the step response.

Simulate the step response of the closed-loop system (VisSim or Simulink preferred with  $K(z)*G(s)$ )

Solution: Use numerical methods to find where the phase is 180 degrees along the damping line:

$$G(s) \cdot ZOH \cdot K(z) = 1 \angle 180^\circ$$

$$\left( \frac{2331}{(s+2.6338)(s+30.2062)(s+53.7896)} \right) \cdot \exp\left(\frac{-sT}{2}\right) \cdot \left(\frac{kz}{z-1}\right) = 1 \angle 180^\circ$$

Searching along the damping line:  $\zeta = 0.4559$

$$s = -1.1363 + j2.2726$$

$$z = e^{sT} = 0.8696 + j0.2011$$

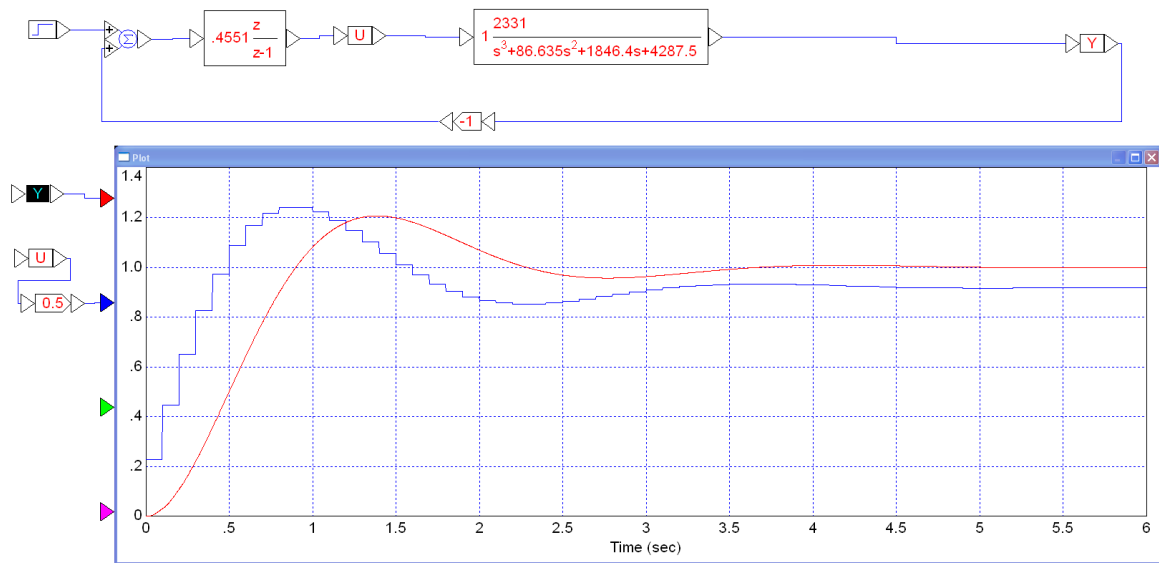
$$\left( \left( \frac{2331}{(s+2.6338)(s+30.2062)(s+53.7896)} \right) \cdot \exp\left(\frac{-sT}{2}\right) \cdot \left(\frac{kz}{z-1}\right) \right)_s = 2.1970k \angle 180^\circ$$

$$k = \frac{1}{2.1970} = 0.4551$$

so

$$K(z) = 0.4551 \left( \frac{z}{z-1} \right)$$

## Checking the step response in VisSim



### Note:

- The overshoot is 20%
- The steady-state error is zero (type-1 system)

2) Assume  $T = 0.1$  seconds and

$$G(s) = \left( \frac{2331}{(s+2.6338)(s+30.2062)(s+53.7896)} \right)$$

Design a digital PI controller

$$K(z) = k \left( \frac{z-a}{z-1} \right)$$

that results in 20% overshoot in the step response. Simulate the step response of the closed-loop system (VisSim or Simulink preferred with  $K(z)*G(s)$ )

Pick 'a' to cancel the pole at  $s = -2.6388$

$$z = e^{sT} = 0.7681$$

The plant + sample & hold + controller is then

$$\left( \frac{2331}{(s+2.6338)(s+30.2062)(s+53.7896)} \right) \cdot \exp\left(\frac{-sT}{2}\right) \cdot k \left( \frac{z-0.7681}{z-1} \right) = 1 \angle 180^\circ$$

Search along the damping line until the angle is 180 degrees

$$s = -5.2219 + j10.4437$$

$$z = e^{sT} = 0.2981 + j0.5129$$

To find k

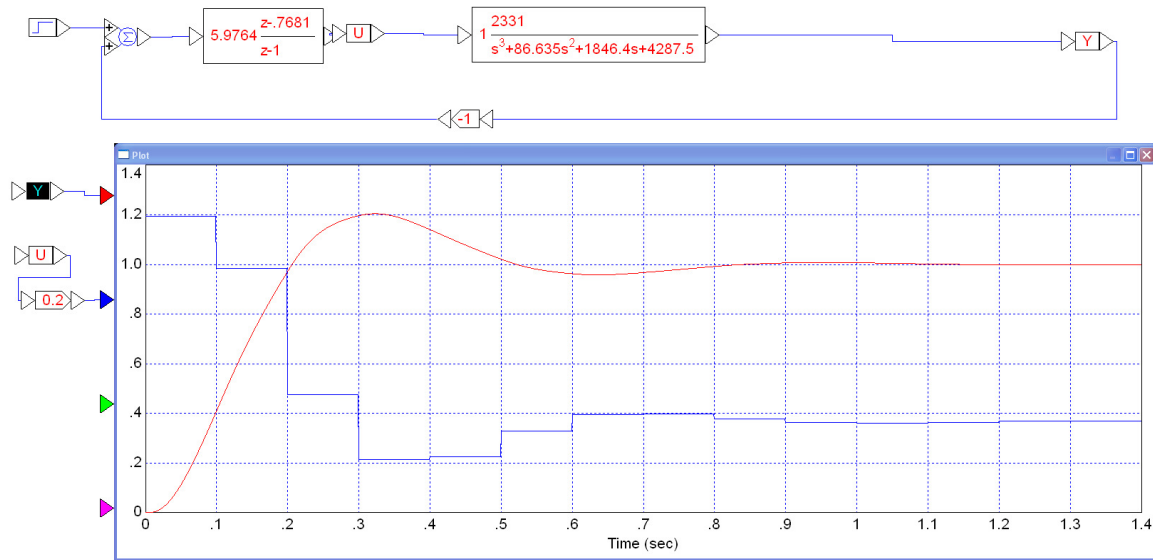
$$\left( \left( \frac{2331}{(s+2.6338)(s+30.2062)(s+53.7896)} \right) \cdot \exp\left(\frac{-sT}{2}\right) \cdot k \left( \frac{z-0.7681}{z-1} \right) \right)_s = 0.1673k \angle 180^\circ$$

$$k = \frac{1}{0.1673} = 5.9764$$

and

$$K(z) = 5.9764 \left( \frac{z-0.7681}{z-1} \right)$$

## Checking in VisSim



### Note:

- 20% overshoot
- 0.8 second settling time

3) Assume  $T = 0.1$  seconds and

$$G(s) = \left( \frac{2331}{(s+2.6338)(s+30.2062)(s+53.7896)} \right)$$

Design a digital PID controller

$$K(z) = k \left( \frac{(z-a)(z-b)}{z(z-1)} \right)$$

that results in 20% overshoot in the step response.

Simulate the step response of the closed-loop system (VisSim or Simulink preferred with  $K(z)*G(s)$ )

Pick the zeros to cancel the poles at  $\{-2.6448, -30.2062\}$

$$s = -2.6448 \quad z = e^{sT} = 0.7676$$

$$s = -30.2062 \quad z = e^{sT} = 0.0488$$

$K(z)$  is then of the form

$$K(z) = k \left( \frac{(z-0.7676)(z-0.0488)}{z(z-1)} \right)$$

*note: This is almost pole-zero cancellation. This is telling you that you need to use a faster sampling rate for a PID controller to have much improvement over a PI)*

The plant + sample & hold + controller is then

$$\left( \frac{2331}{(s+2.6338)(s+30.2062)(s+53.7896)} \right) \cdot \exp\left(\frac{-sT}{2}\right) \cdot k \left( \frac{(z-0.7676)(z-0.0488)}{z(z-1)} \right) = 1 \angle 180$$

Search along the damping line until the angle is 180 degrees

$$s = -5.5801 + j11.1601$$

$$z = e^{sT} = 0.2514 + j0.5142$$

At this point

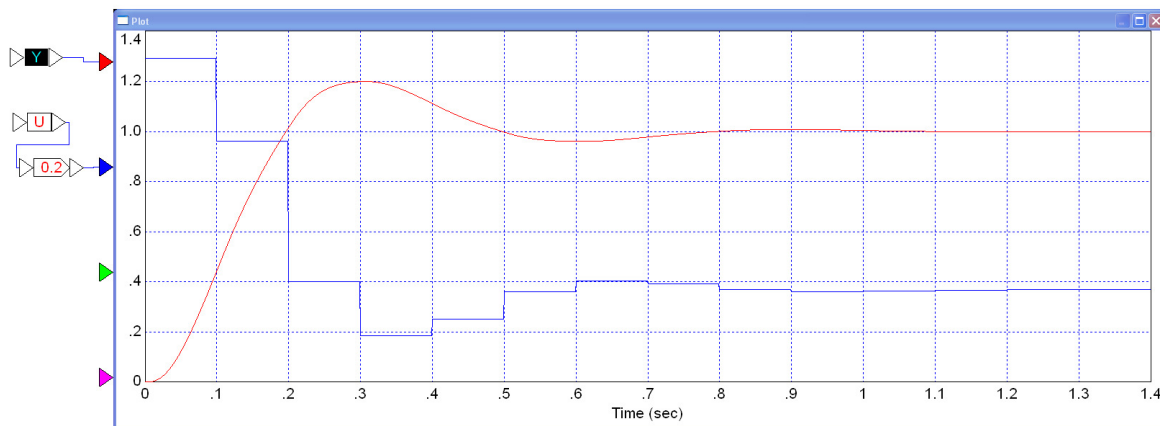
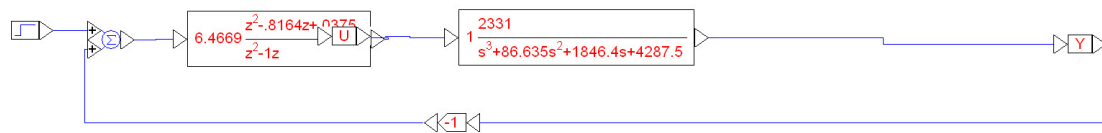
$$\left( \left( \frac{2331}{(s+2.6338)(s+30.2062)(s+53.7896)} \right) \cdot \exp\left(\frac{-sT}{2}\right) \cdot k \left( \frac{(z-0.7676)(z-0.0488)}{z(z-1)} \right) \right)_s = -0.1546k$$

To find  $k$

$$k = \frac{1}{0.1546} = 6.4669$$

so

$$K(z) = 6.4669 \left( \frac{(z-0.7676)(z-0.0488)}{z(z-1)} \right)$$



## Meeting Design Specs

4) Assume a sampling rate of  $T = 0.1$  seconds and

$$G(s) = \left( \frac{2331}{(s+2.6338)(s+30.2062)(s+53.7896)} \right)$$

Design a digital controller that results in

- No error for a step input
- 20% overshoot for the step response, and
- A 2% settling time of 1 seconds

Simulate the step response of the closed-loop system (VisSim or Simulink preferred with  $K(z)*G(s)$ )

Translation:

- Make this a type-1 system
- Place the closed-loop dominant pole at  $s = -4 + j8$
- Or in the z-plane:  $z = 0.4670 + j0.4809$

Pick  $K(z)$  to be of the form

$$K(z) = k \left( \frac{(z-0.7676)(z-0.0488)}{(z-1)(z-a)} \right)$$

The open-loop system is then

$$G(s) \cdot ZOH \cdot K(z) = 1 \angle 180^\circ$$

$$\left( \frac{2331}{(s+2.6338)(s+30.2062)(s+53.7896)} \right) \cdot \exp\left(\frac{-sT}{2}\right) \cdot k \left( \frac{(z-0.7676)(z-0.0488)}{(z-1)(z-a)} \right)$$

Analyze what we know

$$\left( \left( \frac{2331}{(s+2.6338)(s+30.2062)(s+53.7896)} \right) \cdot \exp\left(\frac{-sT}{2}\right) \cdot \left( \frac{(z-0.7676)(z-0.0488)}{(z-1)} \right) \right)_{s=-4+j8}$$

$$= 0.1278 \angle -115.6617^\circ$$

For the phase to add up to 180 degrees, the pole at  $z=a$  must provide a phase shift of

$$\angle(z-a) = 64.3383^\circ$$

Using some trig:

$$a = 0.4670 - \frac{0.4809}{\tan(64.3383^\circ)} = 0.2360$$

Analyzing what we know:

$$\left( \left( \frac{2331}{(s+2.6338)(s+30.2062)(s+53.7896)} \right) \cdot \exp\left(\frac{-sT}{2}\right) \cdot \left( \frac{(z-0.7676)(z-0.0488)}{(z-1)(z-0.2360)} \right) \right)_{s=-4+j8}$$

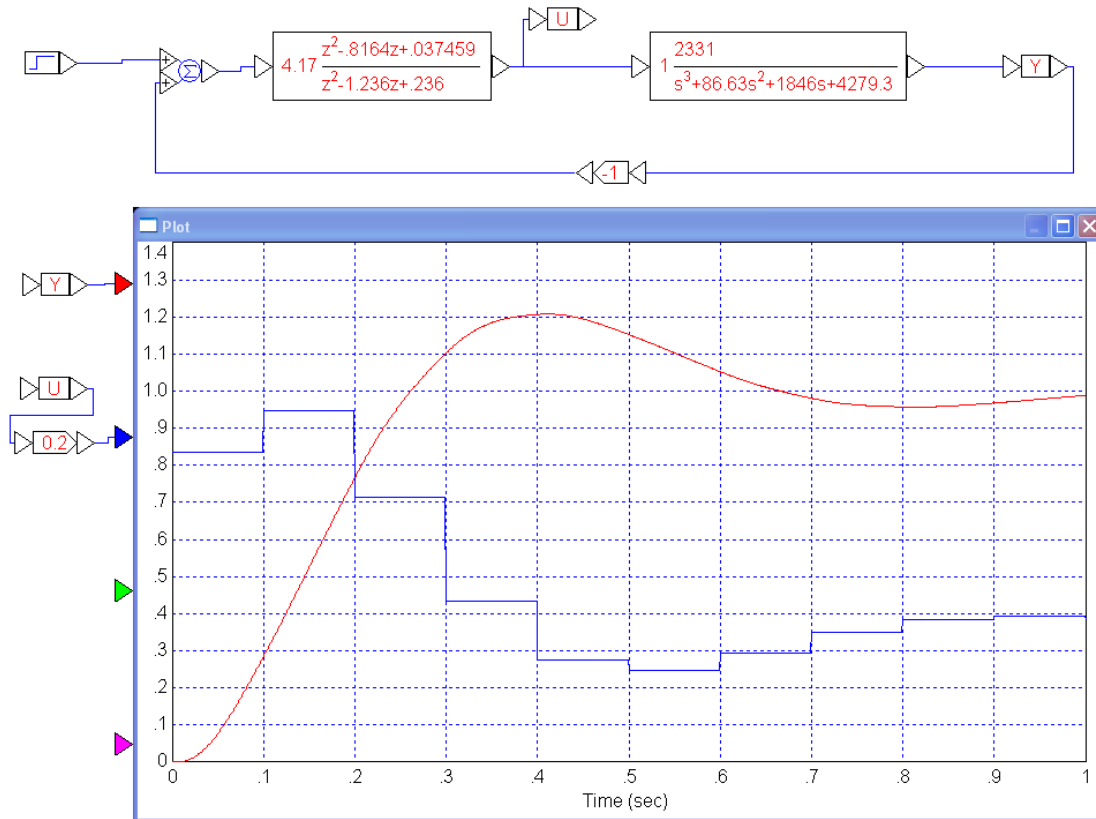
$$= 0.2396 \angle 180^\circ$$

so

$$k = \frac{1}{0.2396} = 4.1737$$

and

$$K(z) = 4.1737 \left( \frac{(z-0.7676)(z-0.0488)}{(z-1)(z-0.2360)} \right)$$





5) Assume

$$G(s) = \left( \frac{2331}{(s+2.6338)(s+30.2062)(s+53.7896)} \right)$$

Design a digital controller with  $T = 0.05$  seconds that results in

- No error for a step input
- 20% overshoot for the step response, and
- A 2% settling time of 1 seconds

Simulate the step response of the closed-loop system (VisSim or Simulink preferred with  $K(z)*G(s)$ )

Note: Changing the sampling rate is a big deal: it means a complete redesign of  $K(z)$

Translation:

- Make this a type-1 system
- Place the closed-loop dominant pole at  $s = -4 + j8$
- Or in the z-plane:  $z = 0.7541 + j0.3188$

Pick  $K(z)$  to be of the form

$$K(z) = k \left( \frac{(z-0.8766)(z-0.2208)}{(z-1)(z-a)} \right)$$

*note: The zeros moved since the sampling rate changed. They're still defined as  $z = \exp(sT)$*

The open-loop system is then

$$\left( \frac{2331}{(s+2.6338)(s+30.2062)(s+53.7896)} \right) \cdot \exp\left(\frac{-sT}{2}\right) \cdot k \left( \frac{(z-0.8766)(z-0.2208)}{(z-1)(z-a)} \right)$$

Analyze what we know

$$\left( \left( \frac{2331}{(s+2.6338)(s+30.2062)(s+53.7896)} \right) \cdot \exp\left(\frac{-sT}{2}\right) \cdot \left( \frac{(z-0.8766)(z-0.2208)}{(z-1)} \right) \right)_{s=-4+j8}$$

$$= 0.1092 \angle -123.01^\circ$$

For the phase to add up to 180 degrees, the pole at  $z=a$  must provide a phase shift of

$$\angle(z-a) = 56.994^\circ$$

Using some trig:

$$a = 0.7541 - \frac{0.3188}{\tan(56.994^\circ)} = 0.5470$$

Analyzing what we know:

$$\left( \left( \frac{2331}{(s+2.6338)(s+30.2062)(s+53.7896)} \right) \cdot \exp\left(\frac{-sT}{2}\right) \cdot \left( \frac{(z-0.8766)(z-0.2208)}{(z-1)(z-0.5470)} \right) \right)_{s=-4+j8}$$

$$= 0.31848 \angle 180^\circ$$

so

$$k = \frac{1}{0.31848} = 3.14$$

and

$$K(z) = 3.14 \left( \frac{(z-0.8766)(z-0.2208)}{(z-1)(z-0.5470)} \right)$$

note:

- The response (red line) is almost unchanged. It should be: it's the same design requirements
- It was a complete redesign for  $K(z)$  due to changing the sampling rate

